Krylov Model-Order Reduction Techniques for Time- and Frequency-Domain Wavefield Problems

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Acknowledgment

This is joint work with

- Mikhail Zaslavsky, Schlumberger-Doll Research
- Jörn Zimmerling, Delft University of Technology

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• Vladimir Druskin, Schlumberger-Doll Research

Outline

- Basic equations and symmetry properties
- Polynomial Krylov reduction 1
- Perfectly matched layers
- Polynomial Krylov reduction 2
- Laurent polynomial (extended) Krylov reduction
- Introducing rational and preconditioned rational Krylov reduction

More on this in J. Zimmerling's talk

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Basic equations

 First-order wavefield system acoustics, seismics, electrodynamics

$$(\mathcal{D}+\mathcal{M}\partial_t)\mathcal{F}=-w(t)\mathcal{Q}$$

- Plus initial conditions
- Dirichlet boundary conditions
- Signature matrix δ^-

$$\delta^- = \operatorname{diag}(1, -1, -1, -1)$$
 (acoustics)

Basic equations

Spatial discretization

$$(D+M\partial_t)f=-w(t)q$$

- Order of this system can be very large especially in 3D
- Discretized counterpart of δ^- is denoted by d^-

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Basic equations

Solution

$$f(t) = -w(t) * \eta(t) \exp(-At)M^{-1}q$$

- $\eta(t)$ Heaviside unit step function
- System matrix

$$A = M^{-1}D$$

Symmetry

- System matrix A is skew-symmetric w.r.t. WM
- Evolution operator exp(-At) is orthogonal w.r.t. WM
- Inner product and norm

$$\langle x, y \rangle = y^H WMx$$
 $||x|| = \langle x, x \rangle^{1/2}$

 Stored field energy in the computational domain (sum of field energies)

$$\frac{1}{2} \|f\|^2$$

• Initial-value problem: norm of f is preserved



- System matrix A is symmetric w.r.t. WMd⁻
- Bilinear form

$$\langle x, y \rangle = y^H W M d^- x$$

• Free field Lagrangian (difference of field energies)

$$\frac{1}{2}\langle f,f
angle$$

• Symmetry property related to reciprocity

Symmetry

Introduce

$$d^{\mathsf{p}} = rac{1}{2}(l+d^{-})$$
 and $d^{\mathsf{m}} = rac{1}{2}(l-d^{-})$

- Write f = f(q) to indicate that the field is generated by a source q
- Reciprocity
- Source vector: $q = d^{p}q$, receiver vector $r = d^{p}r$

$$\langle f(q), r \rangle = \langle q, f(r) \rangle$$

• Source vector: $q = d^{p}q$, receiver vector $r = d^{m}r$

$$\langle f(q), r \rangle = - \langle q, f(r) \rangle$$

Polynomial Krylov reduction

- Exploit symmetry of system matrix in a Lanczos reduction algorithm
- For lossless media both symmetry properties lead to the same reduction algorithm
- First symmetry property is lost for lossy media with a system matrix of the form $A = M^{-1}(D + S)$
- Second symmetry property still holds

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Polynomial Krylov reduction

 For lossless media, FDTD can be written in a similar form as Lanczos algorithm

recurrence relation for FDTD

recurrence relation for Fibonacci polynomials

 Stability of FDTD and numerical dispersion can be studied using this connection

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Polynomial Krylov reduction

- Lanczos recurrence coefficients: β_i
- Comparison with FDTD: $1/\beta_i$ = time step of Lanczos
- Automatic time step adaptation no Courant condition
- Lanczos reduction hardly provides any speedup compared with FDTD
- Both are polynomial field approximations
 - Lanczos: field is approximated by a Lanczos polynomial in A
 - FDTD: field is approximated by a Fibonacci polynomial in A

PML

- No outward wave propagation has been included yet
- Implementation via Perfectly Matched Layers (PML)
- Coordinate stretching (Laplace domain)

$$\partial_k \longleftrightarrow \chi_k^{-1} \partial_k \qquad k = x, y, z$$

Stretching function

$$\chi_k(k,s) = \alpha_k(k) + \frac{\beta_k(k)}{s}$$

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• Stretched first-order system

$$ig[\mathcal{D}(s)+\mathcal{S}+s\mathcal{M}ig]\hat{\mathcal{F}}=-\hat{w}(s)\mathcal{Q}$$

Direct spatial discretization

$$[D(s) + S + sM]\hat{f} = -\hat{w}(s)q$$

• Leads to nonlinear eigenproblems for spatial dimensions > 1

PML

- Linearization of the PML
- Spatial finite-difference discretization using complex PML step sizes

$$(D_{cs}+S+sM) f_{cs}=-w(s)q$$

System matrix

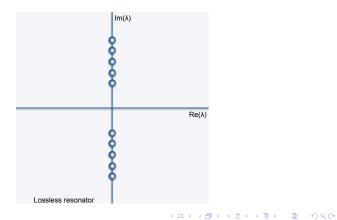
$$A_{\rm cs} = M^{-1}(D_{\rm cs} + S)$$

- V. Druskin and R. F. Remis, "A Krylov stability-corrected coordinate stretching method to simulate wave propagation in unbounded domains," SIAM J. Sci. Comput., Vol. 35, 2013, pp. B376 – B400.
- V. Druskin, S. Güttel, and L. Knizhnerman, "Near-optimal perfectly matched layers for indefinite Helmholtz problems," SIAM Rev. 58-1 (2016), pp. 90 – 116.

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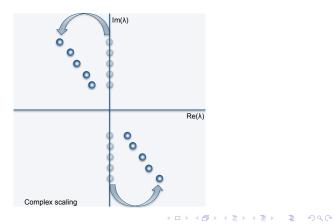
PML

• Spectrum of the system matrix A_{cs}



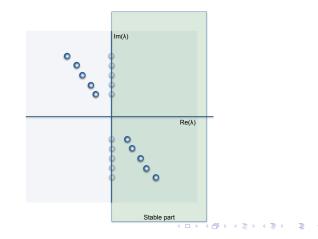
PML

• Eigenvalues move into the complex plane



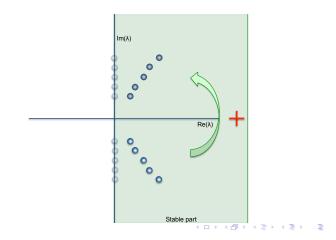
PML

• Stable part of the spectrum



PML

• Stability correction



PML

• Time-domain stability-corrected wave function

$$f(t) = -w(t) * 2\eta(t) \operatorname{Re} \left[\eta(A_{cs}) \exp(-A_{cs}t) q \right]$$

• Complex Heaviside unit step function

$$\eta(z) = egin{cases} 1 & \operatorname{Re}(z) > 0 \ 0 & \operatorname{Re}(z) < 0 \end{cases}$$

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PML

Frequency-domain stability-corrected wave function

$$\hat{f}(s) = -\hat{w}(s) \big[r(A_{\mathsf{cs}},s) + r(\bar{A}_{\mathsf{cs}},s) \big] q$$

with

$$r(z,s)=\frac{\eta(z)}{z+s}$$

Note that \$\hfi(\overline{s}) = \overline{f}(s)\$ and the stability-corrected wave function is a nonentire function of the system matrix \$A_{cs}\$

- Symmetry relations are preserved
- With a step size matrix W that has complex entries
- These entries correspond to PML locations

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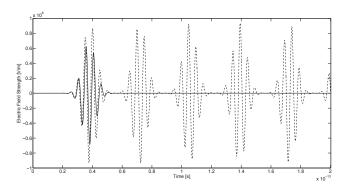
Polynomial Krylov Reduction

- Stability-corrected wave function cannot be computed by FDTD
- SLDM field approximations via modified Lanczos algorithm
- Reduced-order model

$$f_m(t) = -w(t) * 2 \|M^{-1}q\|\eta(t)\operatorname{Re}\left[V_m\eta(H_m)\exp(-H_mt)e_1\right]$$

Polynomial Krylov Reduction

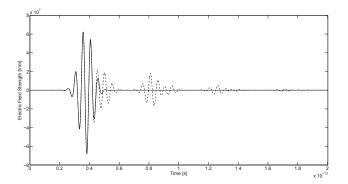
• *m* = 300



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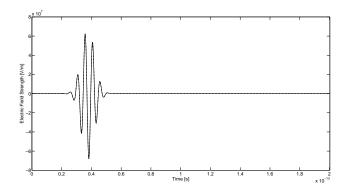
Polynomial Krylov Reduction

• *m* = 400



Polynomial Krylov Reduction

• *m* = 500



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Extended Krylov reduction

- Stability-corrected wave function is approximated by a Lanczos polynomial in A_{cs}
- The wave function is a nonentire function of the system matrix
- Idea: approximate the stability-corrected function by a Laurent polynomial
- Perhaps an even better idea: approximate the stability-corrected wave function by rational functions (more on this later)

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Extended Krylov reduction

Extended Krylov subspace

$$\mathbb{K}_{m_1,m_2} = \operatorname{span}\{\mathsf{A}^{-m_1+1}\mathsf{q},...,\mathsf{A}^{-1}\mathsf{q},\mathsf{q},\mathsf{A}\mathsf{q},...,\mathsf{A}^{m_2-1}\mathsf{q}\}$$

Elements from this space: Laurent polynomials in matrix A acting on the source vector q

Extended Krylov reduction

• Original extended Krylov method of Druskin and Knizhnerman generates the sequence of subspaces

$$\mathbb{K}_{m_1,1} \subset \mathbb{K}_{m_1,2} \subset ... \subset \mathbb{K}_{m_1,m_2}$$

via short-term recurrence relations

- A more general approach was proposed by Jagels and Reichel
- Efficiently generate the sequence of subspaces

$$\mathbb{K}_{1,i+1} \subset \mathbb{K}_{2,2i+1} \subset ... \subset \mathbb{K}_{k,ki+1}$$

again via short term recurrence relations

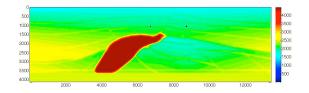
• *i* is an integer

matvec with $A = i \cdot \#$ matvec with A^{-1}

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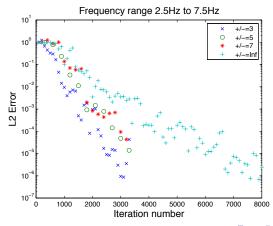
Extended Krylov reduction

• SEG Salt model/velocity profile – 3D acoustics, frequency-domain, order \approx 93 million



Extended Krylov reduction

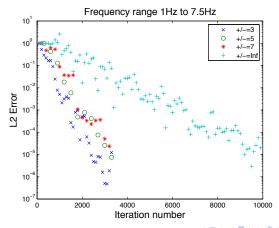
SEG Salt model - generalized EKS implementation



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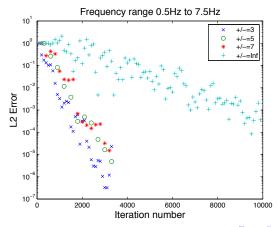
Extended Krylov reduction

• SEG Salt model - generalized EKS implementation



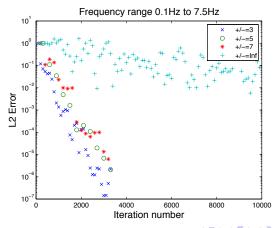
Extended Krylov reduction

SEG Salt model - generalized EKS implementation



Extended Krylov reduction

SEG Salt model - generalized EKS implementation



Rational Krylov reduction

 In a rational Krylov method, we approximate the field by the span of snapshots

$$\hat{f}(s_1), \hat{f}(s_2), ..., \hat{f}(s_m)$$

for different frequencies s_i , i = 1, 2, ..., m

 The snapshots are obtained by solving discretized wavefield systems of the form

$$[D(s_i) + s_i M] \hat{f}(s_i) = -q$$

Rational Krylov reduction

- No PML linearization is necessary!
- When using a rational Krylov method, we deal with the above system directly and not with the stability-corrected system/wave function

Rational Krylov reduction

Return to the system

$$[D(s)+sM]f=-q$$

Symmetry relation

$$D^{T}(s)\tilde{W}(s) = \tilde{W}(s)D(s)$$

Rational Krylov approximations

 ${\ensuremath{\,\circ\,}}$ Multiply by \tilde{W} to obtain

$$A(s)\hat{f}(s)=\tilde{q}$$

• System matrix

$$A(s) = \tilde{W}(s) \left[D(s) + sM \right]$$

Properties

$$A^{\mathcal{T}}(s) = A(s)$$
 and $A^*(s) = A(s^*)$

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Rational Krylov approximations

- Solve system for $m \ge 1$ different frequencies
- Construct the subspace

$$\mathcal{K}_m = \operatorname{span}\{\hat{f}(s_1), \hat{f}(s_2), ..., \hat{f}(s_m)\}$$

and take

$$\mathcal{K}_m^{\mathsf{R}} = \mathsf{span}\{\mathsf{Re}\,\mathcal{K}_m,\mathsf{Im}\,\mathcal{K}_m\}$$

as an expansion and projection space

Note that

$$\hat{f}(s_i) \in \mathcal{K}_m^{\mathsf{R}}$$
 and $\hat{f}(s_i^*) \in \mathcal{K}_m^{\mathsf{R}}$ $i = 1, 2, ..., m$

Rational Krylov approximations

- Let V_m be a basis matrix of $\mathcal{K}_m^{\mathsf{R}}$
- Field approximation

$$f_m(s) = V_m \hat{a}_m(s)$$

- Expansion coeffcients are determined from Galerkin condition
- Structure-preserving reduced-order model

$$\hat{f}_m(s) = V_m \hat{R}_m^{-1}(s) V_m^T \tilde{q}$$
 $R_m(s) = V_m^T A(s) V_m$

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Rational Krylov approximations

Interpolation properties

$$\hat{f}_m(s_i) = \hat{f}(s_i)$$
 and $\hat{f}_m(s_i^*) = \hat{f}(s_i^*)$ $i = 1, 2, ..., m$

• For coinciding source/receiver pairs

$$\frac{\mathsf{d}}{\mathsf{d}s}\tilde{q}^{\mathsf{T}}\hat{f}_m(s)\Big|_{s=s_i,s_i^*} = \frac{\mathsf{d}}{\mathsf{d}s}\tilde{q}^{\mathsf{T}}\hat{f}(s)\Big|_{s=s_i,s_i^*} \qquad i=1,2,...,m$$

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Rational Krylov approximations

- Large travel times: frequency-domain wavefield highly oscillatory in frequency domain
- Rational Krylov method requires many sampling/interpolation points
- Phase-preconditioned rational Krylov method: factor out the strongly oscillating part using high-frequency asymptotics
- Much more on this in talk of J. Zimmerling

Literature

Complex PML step sizes and stability-correction:

- V. Druskin and R. F. Remis, "A Krylov stability-corrected coordinate stretching method to simulate wave propagation in unbounded domains," SIAM J. Sci. Comput., Vol. 35, 2013, pp. B376 – B400.
- V. Druskin, S. Güttel, and L. Knizhnerman, "Near-optimal perfectly matched layers for indefinite Helmholtz problems," SIAM Rev. 58-1 (2016), pp. 90 – 116.
- J. Zimmerling, L. Wei, P. Urbach, and R. Remis, "A Lanczos model-order reduction technique to efficiently simulate electromagnetic wave propagation in dispersive media," *J. Comp. Phys.*, Vol. 315, 2016, pp. 348 – 362.

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Literature

• Extended Krylov subspace method:

- V. Druskin and L. Knizhnerman, "Extended Krylov subspaces: approximation of the matrix square root and related functions," SIAM J. Matrix Anal. Appl., Vol. 19, 1998, pp. 755 – 771.
- C. Jagels and L. Reichel, "Recursion relations for the extended Krylov subspace method," *Linear Algebra Appl.*, Vol. 434, pp. 1716 1732, 2011.
- V. Druskin, R. Remis, and M. Zaslavsky, "An extended Krylov subspace model-order reduction technique to simulate wave propagation in unbounded domains," *J. Comp. Phys.*, Vol. 272, 2014, pp. 608 – 618.

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• Phase-preconditioned rational Krylov method

 V. Druskin, R. Remis, M. Zaslavsky, and J. Zimmerling, "Compressing large-scale wave propagation via phase-preconditioned rational Krylov subspaces," *to appear on ArXiv*, 2017. See also Jörn Zimmerling's talk.